# SERVICE MANUAL CHASSIS CTU-AA 

## 1. Technical specifications

## CHASSIS CTU - AA

Mains voltage
Power cons. at 220V~
Aerial input impedance Min. aerial input VHF Min. aerial input UHF Max. aerial input VHF/UHF Pull-in range colour sync. Pull-in range horizontal sync. Pull-in range vertical sync. Picture tube range


TV Systems

Indications

VCR programs
Tuning and operating system
UV1315A / IEC (VST)

U1343A / IEC (VST)
Local operating functions
: $220-240 \mathrm{~V} \pm 10 \% \mathrm{AC} ; 50 \mathrm{~Hz}( \pm 5 \%)$
: 35 W (14"), 50W (20"/21"), 3W (Stand-By)
: $75 \Omega$ - coax
: $30 \mu \mathrm{~V}$
: $40 \mu \mathrm{~V}$
: 180 mV
$: \pm 300 \mathrm{~Hz}$
$: \pm 600 \mathrm{~Hz}$
: $\pm 5 \mathrm{~Hz}$
: 14" / 20" / 21"
: Mono: 25^1W (14"), 16^2W (20"/21")
: Stereo : 2x16^ 2x4W (21")+
: PAL BG
: PAL I
: PAL BG / SECAM BG DK
: PAL BG I / SECAM BG L L'
: On screen display (OSD) green and menu
: 1 LED (red in ON and blinking red in stand-by)
: 99
$: \square$ VST
: VHFa: 48-168 MHz
: VHFb: 175 - 447 MHz
: UHF: $455-855 \mathrm{MHz}$
: UHF: $455-855 \mathrm{MHz}$
: Vol/Prog, +, -, contrast, colour, brightness and sharpness (and hue on program AV with NTSC signal).

## 2. Connection facilities

## Euroconector:



## 3. Mechanical instructions

For the main carrier two service positions are possible (3.1).
A: For faultfinding on the component side of the main carrier.
B: For (de) soldering activities on the copper side of the main carrier.
Position A can be reached by first removing the mains cord from it's fixation, then loosen the carrier lips (1) and then pulling the carrier panel (2) for approximately 10 cm .

Position B can be reached from position A after disconnecting the degaussing cable. Put the carrier on the line transformer side.

Fig. 3.1


A


Fig. 3.2

## 4. Oscillograms







${ }_{\text {TP37 }}(\mathbb{1} \rightarrow 8 \mathrm{BV} 2 \mathrm{DC}$
$\stackrel{T}{\mathrm{TP} 38}(\mathbb{1} \longrightarrow 45 \mathrm{~V}$ DC

$\mathrm{TP} 39(1)$
TP 39 C $\mathrm{SVDC}_{\mathrm{OV} \text { DC }}$
$\begin{aligned} & \mathrm{TP} 39 \mathrm{O} \rightarrow \mathrm{OVDC} \\ & \mathrm{TP} 40 \mathrm{O}\end{aligned} \mathrm{OVDC}_{\mathrm{OV}}$
$\mathrm{TP} 41(1) \rightarrow 5 \mathrm{VDC}$

## 5. Print Board Layout



## 6. Electrical Diagram





## 6. Electrical Diagram



## 6. Electrical Diagram



## 6. Electrical Diagram



## 6. Electrical Diagram



## 7.- CIRCUIT DESCRIPTION

### 7.1 SMALL SIGNAL \&MICROCONTROLLER/TEXT(Diagram A)

The small signal is processed by IC 7015. This IC called Ultimate One Chip ( UOC ) also has an embedded microcontroller.
The CTU-AA chassis is designed to accept 2 different ICs in the position 7015, TDA9350 for TXT models and TDA9380 for no TXT. The associated circuitry is the same.
The small signal part includes IF detection, video processing, chroma decoder, RGB, sync processor and sound decoder (mono sets). It is fully controlled by the embedded microcontroller.
The microcontroller contains an specific program that assures all the functions of the appliance, including 2 menus, one to control the set (see Instructions Manual) and another for Service Mode (see Service Instruct. chapter 8.).
For stereo models sound decoder (IC 7101) is controlled also for IC 7015 by I2C bus.
The IC for TXT sets (TDA9350) contains a teletext decoder, including the following functions: TXT on/off, reveal, freeze, temporary cancellation, clock, subcode, zoom, index, flof, page $+/$-, $\mathrm{X} / 26$ and $8 / 30$ packet decoding (station identification and start-up page).


Fig. 7.1 TDA 93XX block diagram

### 7.1.1 IF detection

IF detection is intercarrier type, that means sound and picture are detected in the same circuit (PIF).

- IF amplifier (pins 23, 24): The IF signal coming from pin 11 of the tuner (1001) is filtered by the IF SAW filter (1015) and applied to IF amplifier of IC7015 by means of pin 23 and 24 input. The IF bandpass characteristic is determined by the SAW (Surface Acoustic Wave) filter.
- PLL demodulator ( pins 37, 58, 59): The IF-signal is demodulated with the help of a PLL detector, which one is used to regenerate the IF reference signal. Reference signal is determined by the PLL loopfilter pin 37 and calibrated by microcontroller $x$-tal ( pins 58,59). It is alignment free.
Demodulation is achieved by multiplying reference signal with the incoming IF-signal.
- AGC (pin 27): The IF AGC time constant is internally fixed. The tuner AGC voltage (pin 27) is applied to pin 1 of the tuner and adjusted by microcontroller (see chap. 8.4).
- Video output (pin 38) : This baseband CVBS signal with 2.5 Vpp (sync inclusive) of nominal amplitude, contains the FM intercarrier sound signal. Sound is filtered out by a ceramic trap (1032 or 1033) which frequency can be different depending on the system: 5.5 MHz for Pal BG and 6.0 MHz for Pal I.


### 7.1.2 Sound processor

Mono sets:

- FM demodulation (pins 31, 29): FM sound is filtered from CVBS (pin 38) by a loopfilter (pin 31) and demodulated. Decoupling sound demodulator is made by a capacitor on pin 29.
- De-emphasis and audio out (pin 28): De-emphasis is made by C2102 at pin 28. The signal at this pin is drived to the euroconnector sound output (see Diagram C) by transistor 7104.
- External audio in (pin 35): External audio proceeding of pins 2,6 of euroconnector is applied to this pin. Selection between internal or external is done by an internal switching controlled by I2C (see INT/EXT, chapter 7.6).
- Audio out (pin 44): After a volume control (by I2C), this output is drived to the input $\mathrm{IN}+$ of the final sound amplifier IC7187 (Diagram C).
Stereo sets:
SIF signal (pin 38): Video signal containing sound signal is drived to sound stereo processor IC7101 (see chapter 7.6 and diagram E)


### 7.1.3 Video processing

- Video switches (pins 40, 42): The internal CVBS signal proceeding of pin 38 is now fed to pin 40 of 7015 . External CVBS proceeding from pin 20 of Euroconnector is present on pin 42 of 7015. The IC switches between internal and external are controlled by I2C bus (see INT/EXT, chapter 7.1.8).
The selected CVBS signal is internally supplied to the synchronization, teletext acquisition circuits and the video identification circuit for IF ident.
- Luminance processor: CVBS is also internally applied to luminance processor, which is composed of chrominance trap filter, luminance delay line and peaking circuits. Sharpness control modifies peaking by I2C.


### 7.1.4 Chroma processing

This circuit is an internal (no pins associated) automatic decoder for Pal and NTSC systems commanded by I2C bus.

- ACC and chroma filter: Video signal coming from video switches is supplied to PAL/NTSC chroma bandpass filter via a variable gain amplifier which is controlled by ACC and ACL detection circuits.
The dynamic range of the ACC is 26 dB and detects only the burst amplitude. The ACL is a chroma amplitude detector and is active when the chroma/burst ratio exceeds approximately 3. It ensures that CVBS signal to chroma bandpass filter is limited for large.
- PLL/DCO: Chroma PLL is integrated and it operates during the burstkey period; the DCO (digital controlled oscillator) generates a subcarrier signal Fsc in phaselock with the incoming burst signal. The reference signal for the DCO (Fref.) is derived from the crystal oscillator (1679). Different colour frequencies are internally switched by ${ }^{2} \mathrm{C}$. The hue control rotates the DCO reference phases (HO, H90) from -40 to 40 degrees for NTSC signal via ${ }^{12} \mathrm{C}$ bus.
- PAL/NTSC demodulation: The $0^{a}$ \& $90^{\text {² }}$ signals are supplied to the burst demodulator circuits to obtain ( $B-Y$ ) and (R-Y) respectively. Chroma delay lines used in demodulators are internal.


### 7.1.5 RGB processor

- External RGB inputs (pins 46, 47, 48): RGB inputs coming from euroconnector (see diagram C), are AC coupled (C2291/92/93) and converted internally in YUV signals. Then are switched with internal YUV (YUV SWITCH) by fast blanking.
- Fast blanking external (pin 45): When fast blanking is high external RGB is displayed, only if TV is in external AV (program 0). Fast blanking can switch signals for full screen (by a DC voltage) or for a part of the screen (by a pulse voltage).
- Matrix: After switching, YUV signals are converted to R-Y, G-Y and B-Y in the internal MATRIX circuit. Saturation control is received by I2C bus.
- Black stretcher: Internally the luminance signal is connected to the YUV select circuit and via the input clamps it is supplied to the black stretcher. The black stretcher circuit, fully integrated, extends the grey signal level. This extension is dependent upon the difference between actual black level and the darkest part of the incoming video. RGB adder: RGB signals are obtained in this circuit by adding $R-Y, G-Y$ and $B-Y$ proceeding from matrix and $Y$ proceeding from black stretcher.
- RGB OSD: RGB inputs for OSD and TXT are internally inserted to RGB signals. Fast blanking used is also internal. 45).
- Beam current limiter (pin 49): The beam current limiter circuit functions as an average beam current limiter (BCL) as well as peak white limiter (PWL). Both functions reduce the contrast and brightness of RGB signals.
Contrast reduction begins when Vpin $49<3.1 \mathrm{~V}$ and brightness reduction begins when Vpin $49<1.8 \mathrm{~V}$.
BCL: Average beam current present at pin8 of 5445 (Diagram D) is feedback to pin 49 . When beam current is high, voltage of C2462 is lower, D6462 conducts and Vpin 49 decreases.
PWL: This circuit is an internal detection circuit, which comes into action if the difference between measurement DC level and white level exceeds approximately 2.6 V . Then a current of $200 \mu \mathrm{~A}$ is internally produced to discharge 2298 and Vpin 49 decreases.
- RGB output circuit (pins 51, 52, 53): RGB outputs are drived to RGB amplifier (Diagram B).
- Continuous Cathode Calibration (CCC) (pin 50): This circuit is an auto-tuning loop which stabilizes the black level (offset) as well as the cathode drive level (gain) of each gun. Pulses in 3 consecutive lines R, G and B (pins 51, 52, 53) at the end of frame blanking are used on alternating fields, one field for black level, following for cathode drive level.
Cathode currents of 8 mA for black level and 20 mA for cathode drive level are stabilized. Cathode current is present at pin5 of IC7200 (Diagram B) and measured on pin 50. RGB outputs are adapted to keep cathode currents measured, for example when VG2 voltage is adjusted, DC level of RGB outputs is automatically corrected.
- Warm-up detection circuit (pin 50 ): At the start up a DC voltage of 2.5 V is present in the RGB outputs (pins 51,52 , 53 ). As soon as beam current is detected on pin 50, RGB circuit starts in normal operation. If RGB circuit is damaged or grid 2 is low, the RGB circuit could not start (black picture) due to current is not detected.


### 7.1.6 Horizontal synchro

- Start up: The horizontal oscillator starts is commanded by microcontroller. During start up circuit provides a softer operating horizontal output with a higher frequency.
- Hor. sync. separator: Fully integrated sync. separator with a low pass filter, slicing level at $50 \%$ of the synchronized pulse amplitude.
- Horizontal Phi 1 detector (pin 17): This circuit locks the internal line frequency reference on the CVBS input signal. It is composed of a phase comparator and an internal VCO of $25 \mathrm{MHz}(1600 * 15625)$. The free running frequency is stabilized using the 12 MHz reference of the X -tal oscillator.
Phi 1 detector filter is made by external components at pin 50. Phi 1 time constant is automatically controlled by software for broadcasting signals. For video signals (AV and program 99) constant is always fast to prevent top bending on the screen.
- LBF (pin 34): Line Fly Back input, is obtained by the network R3456/55 (Diagram D) and R3371/T7372.
- Sand Castle (pin 34): Sand Castle output obtained at this pin is used only internally. Levels of sandcastle pulse are 5.3 V for burst detection, 3 V for line blanking and 2 V for frame blanking.
- Horizontal Phi 1 detector (pin 16): The flyback position respect line blanking on TRC cathode is controlled by this circuit. Phi 1 detector filter is made by C2305 pin 16. Horiz. shift can be adjusted by I2C ( see chapter 8)
- HOUT (pin 33): Horizontal output is an open collector which one drives the horizontal driver stage (T7440 diagram D).
7.1.7 Vertical synchro and vertical drive
- Vert. sync. separator: It is an internal integrator to separate frame sync. pulses from CVBS.
- Vertical divider system: The divider system uses a counter that delivers the timing for the vertical ramp generator in the geometry processor. The clock is derived from the horizontal line oscillator. Mode used is automatic $50 / 60 \mathrm{~Hz}$ identification with 50 Hz priority.
- Vertical sawtooth generator (pin 25, 26): A reference current of 100 mA is realized at pin 25 by means of an internal reference voltage ( 3.9 V ) and resistor R3341. This 100 mA reference current is used to derive a
16 mA current to charge C2340 (pin 26) during vertical scan. The charge current can be adjusted with the $I^{2} \mathrm{C}$ bus control, (VERT. SHIFT). The external capacitor is discharged during vertical retrace by the vertical divider system.
- Vertical geometry processor: The saw tooth signal that is derived from the sawtooth generator can be controlled by ${ }^{2} \mathrm{C}$ bus. Control adjustments are: S-CORRECT, VERT.SLOPE, VERT.AMP and VERT.SHIFT (See chapter 8).
- Vert. output stage (pins 21, 22): The vertical geometry processor has a differential current output to drive a DC coupled frame deflection (7401).
- EHT compensation (pin 36): A voltage inversely proportional of beam current, proceeding of pin 8 of line transformer ( 5545 diagram D) is applied to pin 36 to stabilize vertical amplitude from EHT variations.
- Vertical guard input (pin 50): A vertical retrace pulse proceeding from pin 8 of IC7401, is added to cut-off voltage at pin 50 (this pin has 2 separated functions, cut-off function see chapter 7.1.5). If there is not pulse or its level is not correct (should be above $3.65 \mathrm{~V}, 0.8 \mathrm{msec}$.), the set goes to standby mode. (See chapter 7.7.1 Protections).


### 7.1.8 Microcontroller/ Text

Following there is an explanation of the different functions of the microcontroller indicating pins number assigned: - Tuning (pin 4): The unit has a VST (Voltage Synthesized Tuning) system. This system works by tuning to a station on the tuner through a linear variation of the tuning voltage (V-VARI) from 0 V to 33 V applied on pin 2 of the tuner. It is generated on pin 4 of the $\mu \mathrm{C}$ and converted to an adequate level for the tuner using T7605.
While searching, $\mu \mathrm{C}$ are always reading internally AFC (Automatic Frequency Control) and video identification signals. When video signal is identified, $\mu \mathrm{C}$ stops searching and do a fine tuning to reach a right AFC value.

- Control key (pin 5): Pin 5 is activated by a DC voltage. When control keys are not activated, a voltage of 3 V 3 is produced by R3618. If one control key is activated, this pin is connected to ground directly or by a resistor R3665, R3667 (Diagram C), decreasing its voltage.
There are 3 voltage levels depending of the key used: OV for MENU, 1V for P - and 1.9 V for $\mathrm{P}+$.
- Mute output (pin 6): This pin is a 3 state output used to control the sound amplifier (see chapter 7.4). States are 0V, open and Vcc (3.3V).
- INT/EXT input (pin 7): The set can switch to external (AV on the screen) by remote control (selecting program 0 ) or by rise edge at pin 8 of euroconnector (see diagram C). The IC switches internally video and audio to external. In both cases the user can switch to internal changing the channel.
- Band switching (pin 8, 11): There are 2 outputs for band switching pin 11 for VHFI and pin 8 for VHFIII. The $\mu \mathrm{C}$ controls the channel band in the tuner by a voltage of +5 V at the correspondent output. UHF band is controlled by means of transistor 7601. If the set is in VHF, one of the pins 8 or 11 are high and transistor conducts in such a way the UHF voltage is 0 V . If the set is in UHF, pins 8 and 11 are low, transistor is cut and the set is in UHF.
- LED (pin 10): The LED (D6663 diagram C) lights up with a low current when the television set is ON and with a high current when the set is on Standby. While the set is receiving a remote control signal, the led is blinking.
- Picture controls (brightness, contrast, colour, sharpness) are processed internally by I2C bus.
- Sound controls (pins 2, 3):

Mono sets: Sound controls are processed internally by I2C bus
Stereo sets: All information necessary to control stereo functions including sound controls, are processed by a
communication I2C bus between the $\mu \mathrm{c}$ and the stereo decoder (IC 7101 pins 4,5 ).

- Standby (pins 1, 33): When the set is switched to stand by, line stops (pin 33). In addition to this, pin 1 switch to low level to make Power Supply in burst mode (see stand by 7.5.3).
- OSD outputs: The RGB and fast blanking signals used for On-Screen Display (OSD) and also for TXT are internally inserted to RGB outputs (pins 51, 52, 53).
- Oscillator (pins 58, 59): A $12-\mathrm{MHz}$ oscillator is determined by a $12-\mathrm{MHz}$ crystal (1679) between pins 58 and 59.
- P.O.R (pin 60) . : Power on reset (POR) is internal, activated when the set is switched on. If the $\mu \mathrm{C}$ shows abnormal behaviour it is advisable to reset it switching off/on the set. Reset can be done also by a short circuit between pins 60 and 61 .
- RC5 (pin 62): The commands transmitted by the remote control handset are received by infrared receiver (1685) and passed to the microcontroller for decoding.
- EEPROM (pins 63 and 64): The microcontroller is connected to non-volatile memory IC7685 (EEPROM) via bus I2C. The following information are stored in the memory:
- Channel data including tuning voltage and band of all the channels.
- Personal preferences (PP), menu mix and child lock on user menu.
- All settings included on Service Menu.
7.1.9 Supplies and decoupings
- Small signal analog supply (pins 14, 39): The same +8 V supply coming from line deflection is connected to pins 14 and 39 for small signal processing. A separated supply (C2036) feeds video and sound traps circuit.
If +8 V supply is not present at pin 39 or pin 14 during the first 4 seconds after switch ON the TV set, the microprocessor goes to Standby mode (See chapter 7.7 Protections).
- Digital supply (pins 54, 15): A +3V3 supply is used for digital circuit of TV processor (pin 54). A decoupling capacitor for this supply is present at pin 15 (C2301). Pin 54 is used also to supply the $\mu \mathrm{C}$.
- Bandgap decoupling (pin 19): The bandgap circuit provides a very stable and temperature independent reference voltage of 4.0 V which one is used specially in the analogue video processor part.
$-\mu$ C supplies (pins $54,56,61$ ): The $\mu \mathrm{C}$ has several +3 V3 supplies,
- Pin 54: Analogue supply (Oscillator, ADC).
- Pin 56: Digital supply to $\mu \mathrm{C}$ core.
- Pin 61: Supply to all output ports of the $\mu \mathrm{C}$.

When the set is in stand by, the $\mu \mathrm{C}$ switch to stand by mode, only 3 V 3 supplies are present but consumption is reduced.

### 7.2 RGB AMPLIFIERS (diagram B)

The RGB signals available at pins 51,52 and 53 of IC7015 (Diagram A) must be amplified and inverted in order to reach the level required for driving the tube. The integrated circuit IC7200 (TDA6107) is used for this purpose. RGB inputs (pins $2,3,1$ ): These signals are connected to negative inputs of internal operational amplifiers. There is and internal reference voltage of 2.5 V for the amplifiers. Negative feedback of amplifiers are internal resistors. RGB outputs (pins 8, 7, 9): These outputs are applied to the RGB cathodes of the tube. Clamping diodes to +200 V $(6201,6216,6229)$ and 1 K 5 series resistors $(3203,3216,3229)$ are added for flash-over protection.

- Continuous Cathode Calibration (CCC) (pins 50): This circuit is an auto-tuning loop which stabilizes the black level (offset) as well as the cathode drive level (gain) of each gun. Pulses in 3 consecutive lines R, G and B (pins 51, 52, 53 ) at the end of frame blanking are used on alternating fields, one field for black level, following for cathode drive level.
Cathode currents
- Cathode current feedback (pin 5): Cathode current used in Continuous Cathode Calibration circuit (See 7.1.5) is obtained at this pin and feedback to pins 50 of IC7015 (Diagram A)
- White adjustment: Gain and cut-off point of RGB amplifiers can be adjusted by I2C in the service menu (See chapter 8)
- Anti-spot circuit: When the set is switched off, the capacitor 2208 keeps charged and a negative voltage is put at VG1 of the tube.


### 7.3 DEFLECTION (Diagrams A, D)

### 7.3.1 Frame deflection (Diagrams A)

Full bridge vertical deflection output performed by the integrated circuit TD8357J (7401).

- Frame supplies (pins 3,6 ): There are 2 supply voltages, +12 V (pin 3 ) is the main supply and +45 V (pin 6 ) is used to supply the output stage during flyback time.
- Vertical inputs (pins 1, 2): (pins 21, 22): The output currents proceeding from pins 21, 22 of IC7015 are converted to voltages by resistors R3411/12 in such a way that 2 inverted sawtooth are present in the differential inputs of amplifier (pins 1, 2).
- Vertical outputs (pins 4, 7): Vertical output is DC coupled to deflection coil. That means linearity adjustment is not necessary. HF loop stability is achieved by a damping resistor R3413.
- Feedback input (pin 9): A voltage proportional to current deflection is present in R3415/16 and a feedback to pin 9.
- Vertical flyback: This signal is internally added to vertical output (pin 7) using supply voltage at pin 6.
- VGUARD circuit (pin 8): This circuit generates a high level during flyback period, and is used to protect the TV set in case of a vertical failure. (See protections, chapter 7.7)
7.3.2 Line deflection (Diagram D)

The final line transistor is driven by the transformer 5441, whose primary winding is driven by the transistor T7440 connected to the line drive output of IC7015 (pin 33).
The horizontal deflection stage is carried out in a conventional way, with the deflection transistor (T7445) and line transformer (5445).
Beam current info $(\mathrm{BCl})$ is present at C 2460 .
There are the following supply voltages obtained from line transformer (5545):

- Frame supply (+12B):Main frame supply present at capacitor C2455.
- Flyback frame supply (+45): Flyback frame supply present at capacitor C2455.
-+8 V stabilizer: +8 V for small signal is made from +12 V (C2454) by D6474 and T7478
-+5 V stabilizer: And additional voltage of +5 V for small signal is made also from +12 V (C2454) by D6479 and T7479.
- Stereo supply (only stereo models): +12 V from C2454 is drived to 7702 (Diagram E) to stabilize +5 V for stereo decoder (7101.
- FF : The heather voltage is reduced by R3235/38 and 5235 (Diagram B) to obtain 6.3Veff at the CRT .


### 7.4 SOUND AMPLIFIER (Diagram C)

Sound amplifier is a Bridge Tied Load (BTL) amplifier including short circuit protection, mute and stand by mode. IC used can be TDA8941P for 14 " and 17" models, TDA8943SF for 20 " and 21 " models or TDA8944J for stereo models. Behaviour is the same for 3 models. TDA8944J is a double amplifier, that means 2 supplies, 2 inputs and 2 outputs.

- Supply (Vcc, SVR): Main supply (Vcc) is taken from +11V of Power Supply (C2540 diagram D). The IC creates internally a half supply, present in SVR pin and decoupled by 10uF capacitor (C2182).
- Sound input ( $\mathrm{IN}+$ ): Amplifiers have differential inputs ( $\mathrm{IN}+$, $\mathrm{IN}-$ ). Audio input is connected to $\mathrm{IN}+$ decoupled by 220 nF capacitor (C2181/84) and IN- is decoupled to ground by other 220nF capacitor (C2179/93). To avoid oscillations there is a 1 n 5 (C2183/95) capacitor connected between both inputs.
- Mode input (MODE): This input is commanded by mute signal proceeding from the $\mu \mathrm{C}$ (pin 6 of 7015 diagram A) and has three modes depending of the voltage level:
- Standby mode (Vmode = Vcc): Consumption is very low (used during stand by).

In this case Vmute=0V, T7657 and T7659 are cut, MODE signal is high

- Mute mode ( $2.5 \mathrm{~V}<\mathrm{Vmode}<\mathrm{Vcc}$ ): No sound output (used when the set is switched on/off, there is no signal, channel search, change of program, etc.)

In this case pin 6 of 7015 is open T7659 conducts (by resistor divider), T7657 are cut, MODE voltage is $5,5 \mathrm{~V}$.

- Operating mode: (Vmode<0.5V): Sound output present (normal operation).

In this case Vmute=3V3 both transistors conduct, mute signal is low.

- Sound output (OUT+/OUT-): Amplified sound is drived to the loudspeakers.

Mono sets:
Headphones output has been connected in such a way that when headphones are connected, loudspeakers are switched off.
Stereo sets:
Headphones output are coming directly from pins 57 and 58 of stereo decoder IC7101 (see chapter 7.6 and diagram E).

### 7.5 POWER SUPPLY (Diagram D)

Mains isolated switched mode power supply (SMPS). controlled in variable frequency mode.
IC7514 (KA5Q075RT) includes control circuit and output MOSFET. Output feedback is done by means of an optocoupler (7515).

- Switching behaviour: The switching period is divided in on-time, when energy is extracted from the mains into the primary winding (8-12 of 5525), off-time, when energy in the transformer is supplied to the loads via secondary windings of 5525 and dead-time when no energy is extracted or supplied.
- Standby mode: Output voltages are present when the set is in stand by. Standby is done cutting line deflection and making the supply in burst mode in such a way that power consumption is very low.


### 7.5.1 Degaussing and Rectifier

- Degaussing: R3501 is a dual PTC (2 PTC's in one housing). After switch on set, PTC is cold so low-ohmic and so degaussing current is very high. After degaussing, PTC is heated so high-ohmic, so in normal operation degaussing current is very low.
- Rectifier: Mains voltage is filtered by L5500, full wave rectified by diodes D6502-D6505 and smoothed by C2505 (300V DC for 220V AC mains).


### 7.5.2 Regulation

- Power MOSFET (pin 1): Power stage is done by an internal MOSFET between pin1 and ground. Current is sensed by an internal resistor and limited to 5A peak. (See chapter 7.7.2 Protections)
- Start up and supply (pin 3): When the set is switched on, a current via R3527 and R3528 is applied to pin 3. When C2514 is charged to 15 V , the power supply starts and the internal MOSFET conducts. A voltage across transformer windings is built up. The voltage across winding 4-2 is rectified by diodes D6523/25 and used to supply the IC on pin 3.

Voltage of pin3 depends on the output voltage and is about 28 V . If Vpin3 is lower than 9 V supply stops and if it is higher than 32V, output voltage is limited. (See chapter 7.7.2 Protections)

- Sync (pin 5): A pulse voltage proceeding from winding 4-2 is applied to this pin. A delay is added by means of C2524, R3522 and R3526 to synchronize starts conduction of MOSFET at minimum Vds voltage.
- Secondary feedback (pin 4): Line supply feedback is obtained at pin 4 by divider R3531/34/19, the differential amplifier (IC7533) and the optocoupler (IC7515). Voltage at this pin is about 1.1 V
- Output regulation: IC7514 stabilizes output voltage by controlling T-on and so the frequency and the duty cycle. Output control is done by the following way:

If line output is higher, current input of the differential amplifier (IC7533) is higher (amplifier has a reference voltage of 2.5 V ).
Current in IC7533 and also in the optocoupler (IC7515) is higher.
Voltage in pin 4 is lower, T-on is internally reduced and output is reduced.
Output voltage of supply can be adjusted by R3518.

### 7.5.4 Secondary side

- Line supply: Line supply present at capacitor C2530 should be adjusted to the correct value (depending on the TRC) by means of P3534.
- Sound supply (+11V): This supply is used for sound output amplifier and to feed 3 V 3 stabilizer.
$-\mu \mathrm{C}$ supply ( +3 V 3 ): +5 V 3 is regulated by T7575, D6570 and T7571.
7.5.3 Stand by:

When the set is put in standby, line driver (pin33 of IC7015, diagram A) stops and deflection supply is not loaded. Only a reduced load to supply the $\mu \mathrm{C}(+3 \mathrm{~V} 3)$ is necessary.
To reduce more the power consumption, during standby T7572 is cut by STBY signal (pin1 of IC7015, diagram A), then current in IC7533 is increased by means of D6531 and output voltages are reduced.
As a consequence of this, the supply switch to burst mode operation. During working time C2514 is charged till 12V, then supply stops and C2514 is discharged till 11V.

### 7.6 STEREO DECODER

The CTU Stereo chassis is based in the TDA9875A chip. This is a single-chip Digital TV Sound Processor for analog and digital multichannel sound systems in TV sets.

In the stereo chassis the sound part in the IC7015 (TDA9350) is disabled. All sound signal, analog and digital is processed in IC7101 (TDA9875A).

The TDA9875A decodes NICAM signals (B/G and I standard) and two-carrier FM modulation A2 signals (B/G standard).

- Control bus (pins 4, 5): Stereo is fully controlled for the $\mu$ C of IC7015 (Diagram A) via I2C-bus. Status information present in internal registers of IC7101 is read by $\mu \mathrm{C}$ to determine whether any action is required. Control is exercised by the $\mu \mathrm{C}$ also, writing in other internal registers of IC7101. This control includes volume, balance, bass, treble, pseudo stereo, spatial, and Automatic Volume Level (AVL).
- Oscillator (pins 18, 19): The circuitry of the crystal oscillator is fully integrated, only a external 24.576 MHz crystal is needed.
- SIF input (pin 12): SIF signal is filtered from video signal coming from pin 38 of IC7015 (Diagram A) by a passband 5.5 MHz to 6.5 Mhz filter (C2740/41/42 and L5741).
- AGC, ADC: SIF is passed through an AGC circuit and then digitized by an 8 -bit ADC operating at 24.576 MHz . The gain of the AGC amplifier is controlled from the ADC output. Then signal is separated in 2 ways, one for FM and other for NICAM.
- FM demodulation: FM signal is fed via a band-limiting filter to a demodulator. If signal is stereo A2, decoder recovers the left and right signal channels from the demodulated sound carriers. A stereo identification signal is sent to the $\mu \mathrm{C}$.
- NICAM demodulation: NICAM signal is transmitted in a DQPSK code at a bit rate of $728 \mathrm{kbit} / \mathrm{s}$. NICAM demodulator performs DQPSK demodulation and feeds the resulting bitstream and clock signal onto the NICAM decoder. A timing loop controls the frequency of the crystal oscillator to lock the sampling rate to NICAM decoder: The device performs all decoding functions in accordance with the "EBU NICAM 728 specification".
- Audio processing: After demodulation, digital signal is selected (FM, A2 or NICAM) and processed according $\mu \mathrm{C}$ control. Following functions are provided: Forced mono, stereo, channel swap, channel 1, channel 2, spatial effects, Automatic Volume Level (AVL), volume, balance, bass and treble.
- DACs (pins 54, 55): Then signals are transformed to analog by DACs and sent to the outputs. Post filter capacitors for DACs are at pins 54 and 55 .
- Sound outputs (pins 60, 61): After a volume control (by I2C), sound outputs (MAIN R and MAIN L) are drived to the inputs $\mathrm{IN}_{+}$(pins 12 and 6) of the final sound amplifier IC7187C (Diagram C).
- Headphone outputs (pins 57, 58): Headphone outputs are drived by followed emitters 7780 and 7781 to the headphones. These outputs are independent of loudspeakers outputs and controls are separated.
- External audio out (pins 47, 48): Signal at these pins are drived directly to the euroconnector sound outputs (see Diagram C).
- External audio in (pins 33. 34): External audio ( R and L ) proceeding of pins 6 and 2 of euroconnector (Diagram C) is applied pins 33 and 34 of IC7101 and digitalized by an ACD circuit. Selection between internal or external is done from the $\mu \mathrm{C}$ (IC7015) by I2C bus (see INT/EXT, chapter 7.6).
- External audio out (pins 47, 48): Signal at these pins are drived directly to the euroconnector sound outputs (see Diagram C).
-+5 V stabilizer: +5 V supply for stereo processor is obtained from +12 V (C2454 diagram D) by D6702 and 7478. This voltage comes to supply voltages of IC1701 (pins 59, 15, 64)
- Digital supplies (pins 15, 64): +5 V supplies at pins 15 and 64 are used for digital circuits of stereo processor. Power ON reset (pin 16): When the set starts, C2754 is slow charged by an internal 50 K resistor to +5 V . When digital supplies are present and oscillator is working, reset is done if pin 16 is low (Vpin16<1.5 V).
- Analog supply voltage (pin 53,59 ): There is a +5 V supply for analog outputs (DACs and operational amplifiers) at pin 59 . A $50 \%$ reference voltage $(2.5 \mathrm{~V}$ ) for these circuits is present at pin 53.
- ADC supply voltage (pins $38,39,40,46$ ) : A 3.3 V supply voltage is obtained internally and decoupled on pin 38 by C2758. Positive and negative reference voltage for ADC is present at pins 39 and 40 . A $50 \%$ reference voltage ( 1.65 V ) for ADC is present at pin 46.
- Front-end supply (pins $7,8,11$ ) : A separated 3.3 V supply voltage (decoupled on pin 7 by C2750) is necessary for SIF demodulator. By means of a 200 mA current generator at pin 8 , a 2 V reference voltage is produced at pin 7 .


### 7.7 PROTECTIONS

If TV does not start or it switches to stand by, try to start again in a few minutes and if problem remains check protection pins explained following.

### 7.7.1 Protections in IC7015 (Diagram A)

- Pins 14, 39:

If +8 V supply is lower than 6 V (pin 39 or pin 14) during the first 4 seconds after switch ON the TV set, the microprocessor goes to Standby mode (See also chapter 7.1.9).

- Pin 50

If the vertical retrace pulse at this pin is not correct (should be above $3.65 \mathrm{~V}, 0.8 \mathrm{msec}$ ) during the first 4 seconds after switch ON the TV set, the vertical guard blanks the RGB outputs and the TV switch to Standby (See also chapter 7.1.7).

This protection operates in the following situations:

- When the vertical deflection fails and there is no pulse.
- When the vertical retrace pulse is higher than 0.9 msec . (for example vertical amplitude is too high)
- An extra protection is obtained when there is too much current in the CRT (for example +200 V supply is missing), because in this case, DC level of pulse is lower by means of cut-off signal. (See fig. 7.7.1)


Fig 7.7.1 Vertical guard protection
To repair the set more easily, Vguard protection can be disabled, in this case the vertical guard blanks the RGB outputs, but TV does not enter in Standby mode. To disable Vguard, item 8 of Service menu MANUAL C-O should be 01 (See chapter 8.1).
Due to a change in MANUAL C-O is difficult to do if the set switch to Standby, Vguard can be disabled also connecting a diode (1N4148) in series with a 2K2 resistor between pins 10 and 7 of IC7015 (Anode at pin 10). Be careful to disconnect this network after the set is repaired.
7.7.2 Protections in IC7514 (Diagram D)

- Pin 1:

Primary current is sensed by an internal resistor for the IC. If this current is increased to 5A peak, supply works in a intermittent way (ON and OFF). That can be provoked for a short circuit on secondary outputs.

- Pin 3:

If Vpin3 is lower than 9 V the under voltage protection stops the supply.
If Vpin3 is higher than 32V over voltage protection limits output voltage.

## 8.- ELECTRICAL ADJUSTMENTS

### 8.1 Service mode

The signal processor IC7015 is fully controlled by I2C, so that the most of adjustments of the set can be made by service menu.

- Enter in Service mode:
- Select program 75 and press at the same time OSD key (+) on RC and MENU key on local key board during 4 seconds.
Service mode is indicated by a S symbol on the down left corner of the screen.
Note : Controls (volume, contrast, brightness and saturation) will be pre-adjusted to the mid position.
- Display Service Menu : When the set is in service mode it is possible to display Service Menu by OSD key ( + ) on RC. Using P+, P-keys of remote control the different items can be displayed (see table 8.1):

| № | ITEM | Value <br> Hexa | № | ITEM | Value <br> Hexa |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  | Description |  |  | Description |  |
| 1 | ADJUST VG2 | PRG | 13 | R GAIN | 20 H |
| 2 | CATH.DRIVE | 0CH | 14 | G GAIN | 20H |
| 3 | TXT V-SHIFT | 02H | 15 | B GAIN | 20 H |
| 4 | TXT H-SHIFT | 02H | 16 | S-CORRECT | 20 H |
| 5 | V-GUARD DISA | PRG | 17 | VERT.SLOPE | 20 H |
| 6 | AGC START | 25H | 18 | HOR.SHIFT | 20H |
| 7 | RESERVED |  | 19 | VERT.AMP | 20 H |
| 8 | MANUAL C-O | 01H | 20 | VERT.SHIFT | 20H |
| 9 | R CUT-OFF | 20 H | 21 | SYS OPTIONS | 00H |
| 10 | G CUT-OFF | 20H | 22 | MENU OPTIONS | 00H |
| 11 | BRIGHTNESS | PP | 23 | READ STATUS | PRG |
| 12 | CONTRAST | PP | 24 | RESERVED | 00H |

Table 8.1. Service menu. Settings are hexadecimal values

- Pre setting values: When E2PROM is replaced, pre-setting values indicated on table 8.1 are stored by the $\mu \mathrm{C}$. (see 8.7 E2PROM).
- Adjust by Service Mode: When a item is selected, using V+, V- keys of remote control it can be adjusted. Items 2, 5, 23 and 24 have fixed values = pre setting values, rest of items see 8.2 to 8.6 .
- Remove service menu: There are 2 ways to remove service menu
- Saving the new settings: Using OSD (+), MENU or INSTALL keys on RC.
- Keeping the old settings: Switching the TV to stand by. Service mode continues active.
- Remove Service Mode: Switching off the TV


### 8.2 Power supply and focusing:

- Power supply voltage:
- Adjust brightness and contrast controls at minimum.
- Connect a DC voltmeter across C2530 (Diagram D).
- Adjust R3534 for a required voltage depending on the model and the TRC used ( see table 11, page 23).
- Focusing:
- Adjust with the potentiometer placed on the line output transformer.


### 8.3 AGC Adjustment

- Connect a pattern generator to the aerial input with RF signal amplitude $=1 \mathrm{mV}$.
- Adjust the value of AGC start (item 6 of Service menu) so that voltage at pin 1 of the Tuner (1001) is 3.7 V .


### 8.4 White D

### 8.4.1 Manual cut-off:

Item 8 of adjusting values should be 00 (see table 8.1), that means set is in automatic cut-off. However it is possible that RGB do not start (black picture), due to grid 2 is not adjusted. In this case we recommend to change to Manual cut-off ( Item $8=01$ ), pre-adjust grid 2 to have a good picture and change to automatic cut-off ( Item $8=0$ ) before continue adjusting (see warm-up detection circuit in 7.1.5)

### 8.4.2 Grid 2:

- Connect a white pattern generator.
- Adjust contrast at 00 and brightness at 36 . ( 00 and 24 in the service menu because hexadecimal values are used
- Put parameter 1 of the service menu ADJUST VG2 = 1. An arrow will be displayed on the left upper corner.
- Adjust VG2 potentiometer (in line transformer ) till the arrow will be transformed to a square.
- In a few seconds, arrow will be disappear and ADJUST VG2 parameter will be come back to 00 .
8.4.3 White checking:
- Connect pattern generator containing grey scale
- Adjust the set to normal operation and reduce the saturation control to minimum.
- Allow the set to warm up about 10 minutes and check visually if the grey scale has correct colour.
- If not, enter to Service menu and adjust $G$ and $B$ gain (items 14 and 15) until a desired grey is obtained. In the case that adjusting is difficult, start again with the setting values of table 8.1 (items $9,10,13,14,15$ ).


### 8.6 Geometry

- Connect a circle pattern generator with the controls at nominal conditions and enter to service menu.

Adjust to have picture centred in vertical position by service menu item 16.

- Horizontal shift:

Adjust to have picture centred in horizontal position by service menu item 18.

- Vertical amplitude:

Adjust picture height to cover the screen by service menu item 19.
Adjustments only if necessary:

- TXT Shift:

Vertical and horizontal shift of OSD or TXT can be adjusted by items 3 and 4 of service menu.

- Vertical slope:

When item 17 is selected, only upper part of the screen can be seen. Vertical slope is adjusted when centre line of the pattern is just visible.

- Vertical S correction:

S correction can be adjusted by item 20 of service menu.

### 8.7 Options:

The type of chassis is defined by items 21 and 22 of service menu. The following alternatives are available:

| SYSTEM OPTIONS(Item 21) | VALUE | MENU OPTIONS (Item 22) | VALUE |
| :---: | :---: | :---: | :---: |
| PAL B / G | 00 | 16 Languages menu | 00 |
| PAL - I | 01 | English bars menu | 01 |
|  |  | Global menu (without languages). | 03 |

Important note: All the chassis have identification sheet when the chassis type is indicated:
"Cod. service: SXXMXX", where SXX means the option of system and MXX means the option of menu
Example: S01M00 means system =01 (Pall) and menu = 00 (13 Languages menu)
When the chassis or the EEPROM (IC7685) have to be replaced, be careful to keep the same type of chassis, setting correctly the chassis options.

### 8.8 Error messages

The microcomputer also detects errors in circuits connected to the $I^{2} C$ (Inter IC) bus. These error messages are communicated via OSD (On Screen Display) :

| Error message | Error description | Component |
| :---: | :--- | :--- |
| F2 | Eprom communication error | IC7685 |
| F3 | $\mu$ C internal error (RA M) | IC7015 $(\mu \mathrm{C})$ |
| F4 | Internal bus communication | IC7015 |
| F5 | SDA2/ SCL2 bus wrong communication | IC7101 (stereo) |
| No sound menu | SDA2/ SCL2 bus no communication | IC7101 (stereo) |
| F6 | Eprom data error | IC7685 |
| F7 | Not allowed options * | Item 21,22 service <br> menu. |
| F8 | Vertical guard | pin 50, IC7015 |

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## 9. Safety instructions, maintenance instructions, warning and notes

## Safety Instructions for Repairs

1. Safety regulations require that during a repair:

- The set should be connected to the mains via an isolating transformer.
- Safety components, indicated by the symbol $\uparrow$ should be replaced by components identical to the original ones
- When replacing the CRT, safety goggles must be worn.

2. Safety regulations require also that after a repair:

- The set should be returned in its original condition.
- The cabinet should be checked for defects to avoid touching, by the customer, of inner parts.
- The insulation of the mains lead should be checked for external damage.
- The mains lead strain relief should be checked onits function - The cableform and EHT cable are routed correctly and fixed with the mounted cable clamps in order to avoid touching of the CRT, hot components or heat sinks
- The electrical resistance between mains plug and the secondary side is checked. This check can be done as follows:
- Unplug the mains cord and connect a wire tween the two pins of the mains plug.
- Switch on the TV with the main switch.
- Measure the resistance value between the pins of the mains plug and the metal shielding of the tuner or the aerial connection on the set. The reading should be between 4.5 M and 12 M .
- Switch off the TV and remove the wire between the two pins of the mains plug.
- Thermally loaded solder joints should be oldered.
-This includes components like LOT, the line utput transistor, flyback capacitor.


## Maintenance Instructions

It is recommended to have a maintenance inspection carried out periodically by a qualified service employee.
The interval depends on the usage conditions.

- When the set is used in a living room the recommended interval is 3 to 5 years. When the set is used in the kitchen or garage this interval is 1 year.
- During the maintenance inspection the above mentioned "safety instructions for repair" should be caried out. The power supply and deflection circuitry on the chassis, the CRT panel and the neck of the CRT should be cleaned.


## Warnings

1.In order to prevent damage to IC's and transistors any flash-over of the EHT should be avoided. To prevent damage to the picture tube the method, indicated in Fig. 9, has to be applied to discharge the picture tube.
Make use of an EHT probe and a universal meter is 0 V (after approx 30s).
2. ESD.

All IC's and many other semi-conductors are sensitive to electrostatic discharges (ESD). Careless handing during repair can reduce life drastically. When repairing, make sure that you are connected with the same potential as the mass of the set via wrist wrap with resistance. Keep components and tools on the same potential.
3. Proceed with care when testing the EHT section and the picture tube.
4. Never replace any modules or any other parts while the set is switched on.
5. Use plastic instead of metal alignment tools. This will prevent any short circuits and the danger of a circuit becoming unstable.
6. Upon a repair of a transistor or an IC assembly (e.g. a transistor or IC with heatsink and spring) remounting should be carried out in the following order:

1. Mount transistor or IC on heatsink with spring.
2. Resolder the joints.

## Notes

1. After replacing the microcomputer first solder the shielding before testing the set. This is needed as the shielding is used for earth connection. If this is not done the set can switch into protection mode (see description of the SMPS).
2. Do not use heatsink as earth reference.
3. The direct voltages and waveforms should be measured relative to the nearest earthing point on the printed circuit board.
4. Voltages and oscillograms in the power supply section have been measured for both normal operation (O) and in the stand-by mode ( $\mathbf{O}$ ). As an input signal a colour bar pattern has been used.
5. The picTure tube PWB has printed spark gaps. Each spark gap is connected between and electrode of the picture tube and the Aguadog coating.


FIG. 9


[^0]:    * If SYS OPTIONS or MENU OPTIONS are not possible in this chassis, the $\mu \mathrm{C}$ change to allowed options and first time TV set is switch on F7 message is displayed. Options should be checked ( see 8.7)

